# Transformer Driver for Isolated RS-485 Interface 

## General Description

The MAX253 monolithic oscillator/power-driver is specifically designed to provide isolated power for an isolated RS-485 or RS-232 data interface. The device drives a center-tapped transformer primary from a 5 V or 3.3V DC power supply. The secondary can be wound to provide any isolated voltage needed at power levels up to 1 W .
The MAX253 consists of a CMOS oscillator driving a pair of N -channel power switches. The oscillator runs at double the output frequency, driving a toggle flip-flop to ensure $50 \%$ duty cycle to each of the switches. Internal delays are arranged to ensure break-beforemake action between the two switches.
The SD pin puts the entire device into a low-power shutdown state, disabling both the power switches and oscillator.

## Applications

Isolated RS-485/RS-232 Power-Supply Transformer Driver
High Noise-Immunity Communications Interface Isolated and/or High-Voltage Power Supplies
Bridge Ground Differentials
Medical Equipment
Process Control
Typical Operating Circuit

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## Transformer Driver for Isolated RS-485 Interface

## ABSOLUTE MAXIMUM RATINGS

| Supply Voltage (VCC) |  |
| :---: | :---: |
| Control Input Voltages (SD, FS) ................-0.3V to (VCC +0.3 V ) |  |
| Output Switch Voltage (D1, D2) | 12V |
| Peak Output Switch Current (D1, D2) | 1A |
| Average Output Switch Current (D1, D2) | 200mA |
| Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ ) |  |
| Plastic DIP (derate 9.09mW/ ${ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) | . 727 mW |
| SO (derate $5.88 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) | 471 mW |
| $\mu \mathrm{MAX}$ (derate $4.10 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) | 330 mW |
| CERDIP (derate $8.00 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) | 640 mW |


| Operating Temperature Ranges |  |
| :---: | :---: |
| MAX253C | .$^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ |
| MAX253E | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| MAX253MJA | . $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Junction Temperatures |  |
| MAX253C_JE_ | $+150^{\circ} \mathrm{C}$ |
| MAX253MJA ..................................................... $+175^{\circ} \mathrm{C}$ |  |
| Storage Temperature Range .......................... $-65^{\circ} \mathrm{C}$ to $+160^{\circ} \mathrm{C}$ |  |
| Lead Temperature (soldering, 10s) .............................. $300^{\circ} \mathrm{C}$ |  |
| Soldering Temperature (reflow) |  |
| PDIP, SO, $\mu \mathrm{MAX}$ lead(Pb)-free ..................................... 260 <br> PDIP, SO, $\mu \mathrm{MAX}$, CERDIP containing lead(Pb) ............ +240 |  |
|  |  |

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}\right.$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Switch On-Resistance | D1, D2; 100mA |  | 1.5 | 4.0 | $\Omega$ |
| Switch Frequency | FS $=$ VCC or open | 250 | 350 | 500 | kHz |
|  | $\mathrm{V}_{\mathrm{FS}}=0 \mathrm{~V}$ | 150 | 200 | 300 |  |
| Operating Supply Current (Note 1) | No load, VSD $=0 \mathrm{~V}$, FS low |  | 0.45 | 5.0 | mA |
| Shutdown Supply Current (Note 2) | SD $=\mathrm{V}_{C C}$ | 0.4 |  |  | $\mu \mathrm{A}$ |
| Shutdown Input Threshold | High | 2.4 |  |  | V |
|  | Low |  |  | 0.8 | $\mu \mathrm{A}$ |
| Shutdown Input Leakage Current |  | 10 |  |  | pA |
| FS Input Threshold | High | 2.4 |  |  | V |
|  | Low |  |  | 0.8 |  |
| FS Input Leakage Current | $\mathrm{V}_{\mathrm{FS}}=0 \mathrm{~V}$ |  |  | 50 | $\mu \mathrm{A}$ |
|  | $\mathrm{FS}=\mathrm{V}_{\text {CC }}$ | 10 |  |  | pA |
| Start-Up Voltage |  | 2.5 | 2.2 |  | V |

Note 1: Operating supply current is the current used by the MAX253 only, not including load current.
Note 2: Shutdown supply current includes output switch-leakage currents.

MAX253

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## Typical Operating Characteristics

(Circuit of Figure $6, \mathrm{~V}_{\mathbb{I}}=5 \mathrm{~V} \pm 10 \%, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


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Typical Operating Characteristics (continued)
(Circuit of Figure 6, $\mathrm{V}_{\mathrm{V}}=5 \mathrm{~V} \pm 10 \%, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


TIME FROM SHUTDOWN TO POWER-UP


CIRCUIT OF FIGURE 6

# Transformer Driver for Isolated RS-485 Interface 

## Pin Configuration

| TOP VIEW |  |
| :---: | :---: |
|  | $\square$ <br> 8 <br> D2 $\square$ GND2 $\square$ $V_{C C}$ <br> 5 <br> N.C. |

Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :--- |
| 1 | D1 | Open drain of N-channel transformer drive 1. |
| 2 | GND1 | Ground. Connect both GND1 and GND2 <br> to ground. |
| 3 | FS | Frequency switch. If FS $=\mathrm{VCC}$ or open, <br> switch frequency $=350 \mathrm{kHz}$; if $\mathrm{V}_{\mathrm{FS}}=0 \mathrm{~V}$, <br> switch frequency $=200 \mathrm{kHz}$. |
| 4 | SD | Shutdown. Ground for normal operation, <br> connect high for shutdown. |
| 5 | N.C. | Not internally connected. |
| 6 | VCC | 5V supply voltage. <br> 7 <br> GND2Ground. Connect both GND1 and GND2 <br> to ground. |
| 8 | D2 | Open drain of N-channel transformer drive 2. |

## Transformer Driver for Isolated RS-485 Interface



Figure 1. Test Circuit


Figure 2. Block Diagram

## Detailed Description

The MAX253 is an isolated power-supply transformer driver specifically designed to form the heart of a fully isolated RS-485 data interface. Completely isolated communications are obtained by combining the MAX253 with a linear regulator, a center-tapped transformer, optocouplers, and the appropriate Maxim interface product (as described in the Isolated RS-485/RS232 Data Interface section).
The MAX253 consists of an RC oscillator followed by a toggle flip-flop, which generates two 50\% duty-cycle square waves, out-of-phase at half the oscillator fre-
quency (Figure 2). These two signals drive the groundreferenced output switches. Internal delays ensure break-before-make action between the two switches.
Ground SD for normal operation. When high, SD disables all internal circuitry, including the oscillator and both power switches.
Pulling FS low reduces the oscillator frequency and lowers the supply current (see Supply Current vs. Temperature in the Typical Operating Characteristics). FS includes a weak pull-up, so it will be set to the highfrequency state if not connected.

## Transformer Driver for Isolated RS-485 Interface



Figure 3. Typical RS-485 Application Circuit, 5V Configuration

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Figure 4. Typical RS-485 Application Circuit, 3.3V Configuration

## Transformer Driver for Isolated RS-485 Interface



Figure 5. Typical RS-232 Application Circuit

# Transformer Driver for Isolated RS-485 Interface 

## Applications Information

Figures 3-5 are typical isolated RS-485/RS-232 data-interface circuits. These circuits withstand 1800VRMS (1sec) and are intended for industrial communications and contro applications where very high voltage transients, differential ground potentials, or high noise may be encountered.

Table 2 lists transformer characteristics for the applications of Figures 3-10. Some suggested manufacturers of transformers, transformer cores, and optocouplers are listed in Table 3, along with their respective phone and fax numbers.

Important layout considerations include:

- For maximum isolation, the "isolation barrier" should not be breached. Connections and components from one side should not be located near those of the other side.
- Since the optocoupler outputs are relatively highimpedance nodes, they should be located as close as possible to the Maxim interface IC. This minimizes stray capacitance and maximizes data rate.
Refer to the $\mu \mathrm{MAX}$ package information for pin spacing and physical dimensions.


## Isolated RS-485 Data Interface

The MAX253 power-supply transformer driver is designed specifically for isolated RS-485 data-interface applications. The application circuits of Figures 3 and 4 combine the MAX253 with a low-dropout linear regulator, a transformer, several high-speed optocouplers, and a Maxim RS-485 interface device. With a few modifications to these circuits, full-duplex communications can be implemented by substituting the MAX481/MAX485 with the MAX490/MAX491 (for data rates up to 2.5Mbps) or substituting the MAX483/MAX487 with the MAX488/MAX489 (for data rates up to 250kbps).
The data transfer rates of the application circuits in Figures 3 and 4 are critically limited by the optocouplers. Table 1 lists suggested optocouplers and the
appropriate Maxim interface device for data-transfer rates up to 2.5 Mbps .

Refer to the MAX1480 data sheet for a complete isolated RS-485 solution in one package.

## Isolated RS-232 Data Interface

The MAX253 is ideal for isolated RS-232 data-interface applications requiring more than four transceivers. The 1W power output capability of the MAX253 enables it to drive more than 10 transceivers simultaneously. Figure 5 shows the typical application circuit for a complete 120kbps isolated RS-232 data interface. The figure also shows how the Sharp PC417 optocouplers can be replaced by the lower-cost 4N25 devices to achieve data-transfer rates up to 9.6 kbps .
For 3.3V operation, substitute the primary portion of Figure 5 with the circuit of Figure 7
For applications requiring two transceivers or fewer, refer to the MAX250/MAX251 or MAX252 data sheet.

Isolated Power Supplies
The MAX253 is a versatile isolated power driver, capable of driving a center-tapped transformer primary from a 5 V or a 3.3V DC power supply (Figures 6 and 7 ). The secondary can be wound to provide any isolated voltage needed at power levels up to 1 W with a 5 V supply, or 600 mW with a 3.3 V supply. Figure 6 shows a typical 5 V to isolated 5 V application circuit that delivers up to 200 mA of isolated 5V power.
In Figure 7, the MAX253 is configured to operate from a 3.3 V supply, deriving a "boost" Vcc for the MAX253 by connecting diodes to both ends of the transformer primary. This produces nearly double the input supply, and may be useful for other applications, as shown in Figure 4. The average current in each MAX253 switch must still be limited to less than 200mA, so the total power available is approximately 600 mW .

Table 1. Optocouplers and RS-485 Interface ICs for Various Data Rates

| DATA RATE | FULL DUPLEX <br> RS-485 IC | HALF DUPLEX <br> RS-485 IC | OPTOCOUPLER <br> FOR DI/RO | OPTOCOUPLER <br> FOR DE |
| :---: | :---: | :---: | :---: | :---: |
| 250 kbps | MAX488/MAX489 | MAX483/MAX487 | PC417* | PC357T* $^{*}$ |
| 2.5 Mbps | MAX490/MAX491 | MAX481/MAX485 | PC410* | PC357T |

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Figure 6. 5V to Isolated 5V Application Circuit


Figure 7. 3.3 V to Isolated 5V Application Circuit

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Figure 8. Typical 4mA to 20 mA Application Circuit

## Output-Ripple Filtering

A simple lowpass pi-filter (Figures 6 and 7 ) can be added to the output to reduce output ripple noise to approximately $10 \mathrm{mVp}-\mathrm{p}$. The cutoff frequency shown is 21 kHz . Since the filter inductor is in series with the circuit output, minimize its resistance so the voltage drop across it is not excessive.

## Isolated $\mathbf{4 m A}$ to 20mA Analog Interface

The 4 mA to 20 mA current loop is a standard analog signal range that is widely used in the process-control industry for transducer and actuator control signals. These signals are commonly referred to a distant ground that may be at a considerably higher voltage with respect to the local ground.
An analog signal in the range of 0.1 V to 0.5 V is applied to the first MAX480 to generate a signal current in the range of $20 \mu \mathrm{~A}$ to $100 \mu \mathrm{~A}$. This low-level signal is transferred across the barrier by the Siemens IL300 linear optocoupler. This device is unique in that it corrects the dominant nonlinearity present in most optocou-
plers-the LED efficiency variation. The IL300 is really two optocouplers in the same package sharing the same LED; one detector is across the isolation barrier, the other is on the same side as the LED (Figure 8). The latter detector is used to generate a feedback signal identical to the signal on the isolated side of the barrier. The current signal transferred across the barrier is converted back to a voltage that matches the input in the 100 mV to 500 mV range. This voltage is then transformed to the final 4 mA to 20 mA current signal range by the second MAX480, Darlington stage, and the $20 \Omega$ resistor.

Isolated ADC
Almost any serial-interface device is a candidate for operation across an isolation barrier; Figure 10 illustrates one example. The MAX176 analog-to-digital converter (ADC) operates from 5 V and -12 V supplies, provided by the multiple-tapped secondary and linear regulators. If some additional isolated power is needed for signal conditioning, multiplexing, or possibly for a

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Figure 9a. Half-Wave Rectifier-Bipolar


Figure 9b. Full-Wave Rectifier-Bipolar


Figure 9c. Full-Wave Rectifier-Unipolar


Figure 10. Typical Isolated ADC Application

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sensor, an extra several hundred milliwatts could easily be supplied by the circuit, as shown. A 12 V supply could be generated by adding two more diodes to the ends of the secondary, and a -5 V supply could be generated by connecting additional diodes to the $1 / 4$ and $3 / 4$ tap points on the secondary. For 5 V only applications, the MAX187 is recommended.

## Component Selection

## Transformer Selection

The transformer primary used with the MAX253 must be a center-tapped winding with sufficient ET product to prevent saturation at the worst-case lowest selected frequency. The MAX253's guaranteed minimum frequency with the FS pin held low is 150 kHz , equating to a maximum period of $6.67 \mu \mathrm{~s}$. The required ET product
for half the primary is simply the product of the maximum supply voltage and half the maximum period. With FS connected high, the guaranteed minimum frequency is 250 kHz , giving a maximum period of $4 \mu \mathrm{~s}$.
The secondary winding may or may not be center tapped, depending on the rectifier topology used. The phasing of the secondary winding is not critical. In some applications, multiple secondaries might be required. Half-wave rectification could be used, but is discouraged because it normally adds a DC imbalance to the magnetic flux in the core, reducing the ET product. If the DC load is imbalanced, full-wave rectification is recommended, as shown in Figure 9b.
The transformer turns ratio must be set to provide the minimum required output voltage at the maximum anticipated load with the minimum expected input volt-

Table 2. Typical Transformer Characteristics

| CHARACTERISTIC |  | 5 V to $\pm 10 \mathrm{~V}$ | 5 V to 5V | 3.3V to 5V | 5V to 24V | 5 V to $\pm 5 \mathrm{~V} ; \pm 12 \mathrm{~V}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Figure |  | 9a | 2, 3, 5, 6 | 4, 7 | 8 | 10 |
| Turns Ratio |  | 1CT*:1 | 1CT:1.3CT | 1CT:2.1CT | 1CT:5CT | 1CT:1.5CT:3CT |
| Typical Windings | Primary | 44CT | 44CT | 28CT | 44CT | 44CT |
|  | Secondary | 44 | 56CT | 56CT | 220CT | 66CT, 132CT |
| Primary ET Product | FS Low | 18.3V- $\mu \mathrm{s}$ | 18.3V-us | 12V- $\mu \mathrm{s}$ | 18.3V- $\mu \mathrm{s}$ | 18.3V- $\mu \mathrm{s}$ |
|  | FS High | 11V- $\mu \mathrm{s}$ | 11V- $\mathrm{\mu s}$ | 7.2V- $\mathrm{\mu s}$ | 11V- $\mu \mathrm{s}$ | 11V- $\mu \mathrm{s}$ |

*CT = Center Tapped
Table 3. Transformer, Transformer Core, and Optocoupler Suppliers

| TRANSFORMERS | TRANSFORMER CORES | OPTOCOUPLERS |
| :--- | :--- | :--- |
| BH Electronics <br> Phone: (507) 532-3211 <br> FAX: (507) 532-3705 | Philips Components <br> Phone: (407) 881-3200 <br> FAX: (407) 881-3300 | Quality Technology <br> Phone: (408) 720-1440 <br> FAX: (408) 720-0848 |
| Coilcraft <br> Phone: (708) 639-6400 <br> FAX: (708) 639-1469 | Magnetics Inc. <br> Phone: (412) 282-8282 <br> FAX: (412) 282-6955 | Sharp Electronics <br> Phone: (206) 834-2500 <br> FAX: (206) 834-8903 |
| Coiltronics <br> Phone: (516) 241-7876 <br> FAX: (516) 241-9339 | Fair-Rite Products <br> Phone: (914) 895-2055 <br> FAX: (914) 895-2629 | Siemens Components <br> Phone: (408) 777-4500 <br> FAX: (408) 777-4983 |

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age. In addition, include in the calculations an allowance for worst-case losses in the rectifiers. Since the turns ratio determined in this manner will ordinarily produce a much higher voltage at the secondary under conditions of high input voltage and/or light loading, be careful to prevent an overvoltage condition from occurring (see Output Voltage vs. Load Current in the Typical Operating Characteristics).
Transformers used with the MAX253 will ordinarily be wound on high-permeability magnetic material. To minimize radiated noise, use common closed-magneticpath physical shapes (e.g., pot cores, toroids, E/I/U cores). A typical core is the Philips 213CT050-3B7, which is a toroid 0.190 " in diameter and 0.05 " thick. For operation with this core at 5.5 V maximum supply voltage, the primary should have approximately 22 turns on each side of the center tap, or 44 turns total. This will result in a nominal primary inductance of approximately $832 \mu \mathrm{H}$. The secondary can be scaled to produce the required DC output.

## Diode Selection

The MAX253's high switching frequency demands high-speed rectifiers. Schottky diodes are recommended. Ensure that the Schottky diode average current rating exceeds the load-current level. The 1N5817
is a good choice for through-hole applications, and the NIEC* SB05W05C dual in an SOT-23 package is recommended for surface-mount applications. Use the higher frequency setting to reduce ripple.

## Output Filter Capacitor

In applications sensitive to output-ripple noise, the output filter capacitor C2 should have a low effective series resistance (ESR), and its capacitance should remain fairly constant over temperature. Sprague 595D surface-mount solid tantalum capacitors and Sanyo OS-CON through-hole capacitors are recommended due to their extremely low ESR. Capacitor ESR usually rises at low temperatures, but OS-CON capacitors provide very low ESR below $0^{\circ} \mathrm{C}$.
In applications where output ripple is not critical, a $0.1 \mu \mathrm{~F}$ chip or ceramic capacitor is sufficient. Refer to Table 4 for suggested capacitor suppliers. Use the higher frequency setting to reduce ripple.

## Input Bypass Capacitor

The input bypass capacitor C1 is not critical. Unlike switching regulators, the MAX253's supply current is fairly constant, and is therefore less dependent on the input bypass capacitor. A low-cost $0.1 \mu \mathrm{~F}$ chip or ceramic capacitor is normally sufficient for input bypassing.

Table 4. Suggested Capacitor Suppliers

| PRODUCTION METHOD | CAPACITORS |
| :--- | :--- |
|  | Matsuo <br> 267 series (low ESR) <br> USA Phone: (714) 969-2491, FAX: (714) 960-6492 <br> Sprague Electric Co. <br> Sprface Mount <br> S95D293D series (very low ESR) <br> USA Phone: (603) 224-1961, FAX: (603) 224-1430 <br> Murata Erie <br> Ceramic <br> USA Phone: (800) 831-9172, FAX: (404) 436-3030 |
| High-Performance <br> Through Hole | Sanyo <br> OS-CON series (very low ESR) <br> USA Phone: (619) 661-6835, FAX: (619) 661-1055 <br> Japan Phone: 81-7-2070-1005, FAX: 81-7-2070-1174 |
|  | Nichicon <br> PL series (low ESR) <br> USA Phone: (708) 843-7500, FAX: (708) 843-2798 <br> Japan Phone: 81-7-5231-8461, FAX: 81-7-5256-4158 |

* Nihon Inter Electronics Corp.

USA Phone: (805) 867-2555
FAX: (805) 867-2556
Japan Phone: 81-3-3494-7411
FAX: 81-3-3494-7414

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## Chip Information

PROCESS: CMOS

## Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "\#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | DOCUMENT NO. |
| :---: | :---: | :---: |
| $8 \mu \mathrm{MAX}$ | $\mathrm{U} 8+1$ | $\underline{21-0036}$ |
| 8 PDIP | $\mathrm{P} 8+1$ | $\underline{21-0043}$ |
| 8 SO | $\mathrm{S} 8+4$ | $\underline{21-0041}$ |
| 8 CDIP | $\mathrm{J} 8-2$ | $\underline{21-0045}$ |

# Transformer Driver for <br> Isolated RS-485 Interface 

Revision History

| REVISION <br> NUMBER | REVISION <br> DATE | DESCRIPTION | PAGES <br> CHANGED |
| :---: | :---: | :--- | :---: |
| 0 | $1 / 94$ | Initial release | - |
| 1 | $8 / 09$ | Deleted the MAX253EUA part number from the Ordering Information table | 1 |
| 2 | $4 / 10$ | Added automotive qualified part number to the Ordering Information table | 1 |

maxim
integrated $_{\text {mw }}$

[^1]
[^0]:    * PC-Series Optocouplers, Sharp Electronics

    USA Phone: (206) 834-2500
    FAX: (206) 834-8903
    Sharp Electronics, Europe GmbH Germany Phone: (040) 2376-0
    FAX: (040) 230764

[^1]:    Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

